Floating Gate Dosimeter (FGDOS®)

Target Specification. Preliminary



Rev. A0.11 Page 1/22

FEATURES

FGDOS® radiation sensor with digital output

200 uGy (20 mrad) resolution

Total radiation dose up to 250 Gy (TID)

Chip Serial Number

Interface for microcontroller applications

QFN32 with two independent sensors for redundancy

Programmable Sensitivity 5 kHz/Gy or to 30 kHz/Gy

Passive detection mode (zero power consumption)

Temperature monitor integrated on-chip

5V supply voltage

APPLICATIONS

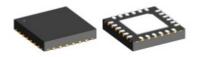
Radiation sensor

Active and Passive Dosimetry

Space

Particle Physics Facilities

PACKAGE



QFN32 5x5m (2 sensors)

GENERAL DESCRIPTION

FGD-02F is a digital radiation sensor based in FGDOS® principle.

Sensor output is a frequency modulated pulse train proportional to radiation dose.

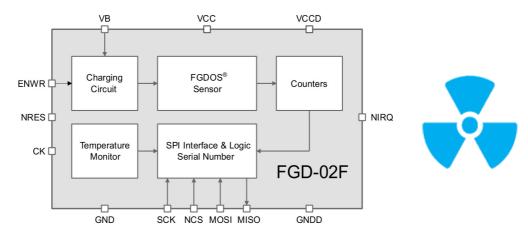
Internal counters allow radiation dose digital value to be read via SPI Interface.

Chip serial number is provided for sensor tracking.

In passive mode, the chip is still sensing the accumulated radiation dose even when there is no power supply.

Sensor temperature dependency is internally compensated.

Additionally, on-chip temperature sensor and reference channel are provided for extended precision applications, via digital post-processing.



BLOCK FUNCTIONAL DIAGRAM



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 2/22

CONTENTS

CONTENTS2	INTERNAL TEN
PACKAGING INFORMATION AND	SERIAL ID NUI
DIMENSIONS	SERIAL PHERI
ABSOLUTE MAXIMUM RATINGS4	(SPI)SPI slave inter
THERMAL DATA4	DATA POST-PR
ELECTRICAL CHARACTERISTICS5	Background Compensating
PRINCIPLE OF OPERATION7	temperature d
QUICK SET-UP AND OPERATION EXAMPLE8	REGISTER MA
BASIC OPERATION OVERVIEW10	DESIGN REVIE
OPERATION DESCRIPTION10	Functions
MEASUREMENT WINDOW SETTING11	DATASHEET R
SENSITIVITY CONFIGURATION12	SOLDERING C
RECHARGING SYSTEM12	DISCLAIMER
POWER SAVING MODE AND PASSIVE DETECTION14	ORDERING INI

INTERNAL TEMPERATURE MONITOR.	14
SERIAL ID NUMBER	14
SERIAL PHERIPHERAL INTERFACE (SPI)	15 15
DATA POST-PROCESSING Background Compensating for the FS and FR temperature dependence	16
REGISTER DESCRIPTION	18
REGISTER MAP	20
DESIGN REVIEW: Notes On Chip Functions	21
DATASHEET REVISION HISTORY	21
SOLDERING CONSIDERATIONS	21
DISCLAIMER	22
ORDERING INFORMATION	22

Floating Gate Dosimeter (FGDOS®)

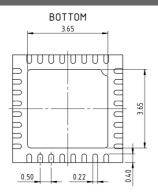


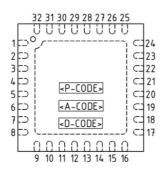
Target Specification. Preliminary

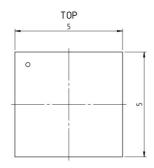
Rev. A0.11 Page 3/22

PACKAGING INFORMATION AND DIMENSIONS









Pin configuration QFN32-5x5 (top view)
The *Thermal Pad* is to be connected to a Ground Plane on the PCB.

Only pin 1 marking on top or bottom defines the package orientation

All dimensions given in mm
Tolerances according to JEDEC MO-220.
The approximated chip weight is 2 g

Pin	Name	Function	Pin	Name	Function
1	MISO-1	Serial Data Output SENSOR-1	17	MISO-2	Serial Data Output SENSOR-2
2	NCS-1	Not Chip Select SENSOR-1	18	NCS-2	Not Chip Select SENSOR-2
3	MOSI-1	Serial Data Input SENSOR-1	19	MOSI-2	Serial Data Input SENSOR-2
4	Rsv.	Connect to GND	20	Rsv.	Connect to GND
5	N.C.	Not Connected	21	N.C.	Not Connected
6	N.C.	Not Connected	22	N.C.	Not Connected
7	NIRQ-1	Not Interrupt Req. SENSOR-1	23	NIRQ-2	Not Interrupt Req. SENSOR-2
8	VCCD-1	Dig. Power Supply SENSOR-1	24	VCCD-2	Dig. Power Supply SENSOR-2
9	GND-1	Ground SENSOR-1	25	GND-2	Ground SENSOR-2
10	VB-1	Recharge Voltage SENSOR-1	26	VB-2	Recharge Voltage SENSOR-2
11	VCC-1	Power Supply SENSOR-1	27	VCC-2	Power Supply SENSOR-2
12	GNDD-1	Digital Ground SENSOR-1	28	GNDD-2	Digital Ground SENSOR-2
13	SCK-1	Serial Clock SENSOR-1	29	SCK-2	Serial Clock SENSOR-2
14	CK-2	Window Clock SENSOR-2	30	CK-1	Window Clock SENSOR-2
15	ENWR-2	Enable Write SENSOR-2	31	ENWR-1	Enable Write SENSOR-1
16	NRES-2	Not Reset SENSOR-2	32	NRES-1	Not Reset SENSOR-1



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 4/22

ABSOLUTE MAXIMUM RATINGS

These ratings do not imply permissible operating conditions; functional operation is not guaranteed. Exceeding these ratings may damage the device

Item	Symbol	Parameter	Conditions			Unit
No.				Min	Max	
G001	VB	Permissible Voltage at VB			20	V
G002	V()	Voltage at NIRQ, VCCD, VB, VCC, SCK, NCS, ENWR, NRES, MISO, NCS, MOSI, CK	Referenced to GND		5.5	V
G003	Vd()	ESD Susceptibility at all pins			TBD(*)	kV
G004	Tj	Junction Temperature		-40	150	°C
G005	Ts	Storage Temperature Range		-40	150	°C

^(*) Electrostatic discharges may vary the charge stored in FGDOS®

THERMAL DATA

These ratings do not imply permissible operating conditions; functional operation is not guaranteed. Exceeding these ratings may damage the device

Item	Symbol	Parameter	Conditions				Unit
No.				Min	Тур	Max	
T01	Та	Operating Ambient Temperature Range		-40		85	°C
T02	Rthja	Thermal Resistance Chip/Ambient	Mounted on PCB		25		K/W
T03	RthjTP	Thermal Resistance Chip/Thermal Pad			4		K/W



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 5/22

ELECTRICAL CHARACTERISTICS

Operating Conditions: VB=18V, VCC=4.5V ... 5.5V, VCCD = VCC, Tj=-40 ... 85 °C, Rad. source = Co60, TID=0Gy unless otherwise stated

stated	stated								
Item No.	Symbol	Parameter	Conditions	Min	Тур	Max	Unit		
Total [Device								
001	VB	Permissible Programmer Voltage at VB	Referenced to GND	15	18	20	V		
002	I(VB)	Recharge current at VB	Recharge Disabled Recharge Enabled			1 50	μΑ		
003	VCC	Permissible Supply Voltage at VCC	Referenced to GND	4.5		5.5	V		
004	I(VCC)	Supply current at VCC	POWR(2:0) = 111 High-sensitivity Mode Low-sensitivity Mode POWR(2:0) = 000			10 5 TBD	mA		
005	VCCD	Permissible Supply Voltage at VCCD	Referenced to GND	4.5		5.5	V		
006	I(VCCD)	Supply current at VCCD	POWR(2:0) = 111 POWR(2:0) = 000			2 TBD	mA		
007	Vc(lo)	Clamp Voltage at VB,VCC, VCCD, MISO, NIRQ, ENWR, NCS, SCK, MOSI, CK	I()=10mA	-1.5		-0.6	V		
008	f(CK)	Recommended CK frequency	ENGATE=0		32.768		kHz		
Digita	I Input/Outp	outs							
100	lsc()lo	Short Circuit Current lo at NIRQ, MISO		-40		-4	mA		
101	Isc()hi	Short Circuit Current hi at NIRQ, MISO		4		40	mA		
102	Vs()lo	Saturation Voltage lo at NIRQ, MISO	I()=2mA	-0.4			V		
103	Vs()hi	Saturation Voltage hi at NIRQ, MISO	I()=-2mA			0.4	V		
104	Vt()hi	Input Threshold Voltage hi at ENWR, NCS, SCK, MOSI, NRES, CK				2	V		
105	Vt()lo	Input Threshold Voltage Io at ENWR, NCS, SCK, MOSI, NRES, CK		0.8			V		
106	I()pd	Pull down Current at ENWR, NCS, SCK, MOSI, NRES. CK		1		50	μА		
Senso	or Output								
200	PSRR()	Power supply rejection Ratio	High Sensitivity Configuration		0.5		Hz/mV		
201	MDD()	Minimum Detectable dose	Constant Temperature, High Sensitivity Configuration		200		μGy		
202	ΔFs()R	Frequency sensitivity	(*) High Sensitivity Configuration Low Sensitivity Configuration		30 5		kHz / Gy		
203	RGmax	Maximum Gamma Dose (TID)			250		Gy		
204	RPMax	Maximum Proton Dose (TID)			200		Gy		
205	ΔFs()An	Annealing Frequency variation after first recharge	(*) Measured 5 Days after		4		kHz		
206	Fs()Noise	Sensor Frequency noise	Constant temperature, no radiation applied		10		Hz		



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 6/22

207	LinG()	Gamma linearity response	Sensor within 20 kHz linear range		99		%
208	LinP()	Protons linearity response	Sensor within 20 kHz linear range		99		%
209	Fs()Temp	Frequency dependency to Temperature	High Sensitivity Configuration Low Sensitivity Configuration		130 500		Hz/°C
Refere	ence Output						
300	Fr()	Reference Frequency	REF(2:0) = 100 Low Sensitivity Configuration High Sensitivity Configuration		130 50		kHz
301	Fr()Noise	Reference Frequency noise	Constant temperature, no radiation applied		10		Hz
302	Fr()Temp	Frequency dependency to Temperature	High Sensitivity Configuration Low Sensitivity Configuration		200 500		Hz/°C
Windo	ow Measurer	ment Gate					
400	Tcklmin	Minimum CK low time between Sensor and Reference measurements	ENGATE=1			20	μs
Temp	erature Moni	itor					
501	Trange	Temperature Measurement Range		-40		125	°C
502	Tresol	Temperature Measurement Resolution			1		°C
503	Reading	Temperature Value Ranges	Tj = 125 °C Tj = -40 °C	200 40		230 60	Digits

^(*) TID and Radiation Dose Rate as specified in attached report



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 7/22

PRINCIPLE OF OPERATION

The **FGDOS**® principle of detection is based on a Floating Gate (FG) capacitor. Charge is pre-stored in the FG using an on-chip recharging system. This charge is stored indefinitely, unless ionizing radiation is applied. When this occurs, the prestored charge at FG discharges. Thus, monitoring the charge at the FG capacitor, radiation dose can be measured.

FGDOS[®] working principle is based on three basic steps, as shown in Figure 1:

- 1. Initial charge action of the FG up to target value (Zone A). In this step, the FG sensor core is evaluated and it is a factory procedure.
- 2. The FG discharges due to applied ionizing radiation (Zone B). The discharging rate of the sensor is highly linear with radiation dose.
- 3. Recharge is triggered when FG charge reaches the threshold value (Zone C).
- 4. The measured radiation dose can be obtained by reading the sensor output data, calculating the sensor value decrease and counting the number of recharges been triggered.

Following these basic steps, **FGDOS**® ensures working in a very linear zone of detection, keeping the charge in the FG between target and threshold value.

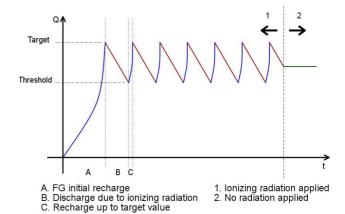


Figure 1. Working principle of FGDOS®

Note: More detailed information on the **FGDOS**® principle of operation, can be found at the following scientific publications:

- S. Danzeca, J. Cesari, M. Brugger, L. Dusseau, A. Masi, A. Pineda, G. Spiezia, "Characterization and Modeling of a Floating Gate Dosimeter with gamma and protons at various energies", November 2014 IEEE Transactions on Nuclear Science, vol. 61, no. 6, pp 3451 – 3457, 2014.
- 2. J. Cesari, A. Barbancho, A. Pineda, G. Ruy and H. Moser "Floating Gate Dosimeter Measurements at 4M Lunar Flyby Mission", The Nuclear and Space Radiation Effects Conference (NSREC) Radiation Effects Data Workshop (REDW), Boston, July 2015.



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 8/22

QUICK SET-UP AND OPERATION EXAMPLE

This section provides, as an example, the steps for a quick set-up of **FGD-02F** working with the most common features:

- One (1) second long Sensor and Reference measurements
- High Sensitivity Mode (HS)
- Automatic recharging with interrupt on new data

These are the required configuration steps:

A) Power-on the device

- 1. Apply a 32.768 kHz clock at **CK** pin.
- 2. Apply supply voltage and wait until NIRQ = 1
- B) Configure FGD-02F for HS mode, 1second long *Measurement Windows*, and Reference Oscillator to 50 kHz nominal frequency.

Addr. 0x0B = b01000000 = 0x40

Addr. 0x0C = b01111001 = 0x79

Addr. 0x0E = b00110000 = 0x30

- C) Configure FGD-02F in Automatic Recharging mode with Interrupt on new data
 - 1. Disconnect Recharging System

Addr. 0x0D = b00000000 = 0x00

- 2. Wait few *Measurement Window* cycles (e.g. 4 cycles) to allow Reference to stabilize.
- 3. Read Reference register **F1R(17:0)**
- 4. Configure **TARGET(7:0)** with the 8 MSB bits from **F1R(17:0)**

Addr. 0x09 = F1R(17:10)

Note

Steps C.2 and C.3 are strictly necessary only for the first time sensor is used. In any other case, microcontroller may have stored typical F1R(17:10) value in order to write it into TARGET(7:0)

5. Configure **THRESHOLD(7:0)** with equivalent 30 kHz nominal value

Addr. 0x0A = b00011101 = 0x1D

Note:

FGD-02F is pre-charged in factory to a nominal, approximated value of 50kHz.

However, it is possible the sensor suffers some discharge during soldering, or accidental exposure to radiation during transportation.

In that cases, it may be desirable to force an initial recharge to ensure initial sensor value to be around 50kHz.

To do so, configure

THRESHOLD(7:0)=TARGET(7:0) for the first time the sensor is used.

This initial recharge operation is finished when register RECHV=0.

Then, configure THRESHOLD(7:0) with equivalent 30 kHz nominal value

6. Enable Interrupt on new data

Addr. 0x0E = b01110000 = 0x70

7. Enable Recharging System

Addr. 0x0D = b01000001 = 0x41



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 9/22

D) Read Measured Data

- 1. When **NIRQ** = 0 read all memory data
- 2. If a recharge is ongoing: **RECHEV** = 1. Disregard *Sensor* value. Go back to point D.1 and wait for next data.

E) Calculate Radiation value

- 1. Convert F1S(17:0) to frequency
- 2. Calculate radiation measured.

$$Radiation = \frac{f\left(Sensor\right)n - f\left(Sensor\right)n - 1}{Sensitivity}$$

 In case it is necessary, apply temperature compensation (see section DATA POST-PROCESSING)



Floating Gate Dosimeter (FGDOS®)

SealiconThe power of small things

Target Specification. Preliminary

Rev. A0.11 Page 10/22

BASIC OPERATION OVERVIEW

FGD-02F contains an **FGDOS**® sensor (*Sensor*) that discharges with radiation dose. The output is encoded in frequency and it decreases as the sensor discharges.

When the sensor value goes below a predefined value, it is possible to recharge the sensor and continue with the radiation dose measurement (see RECHARGING SYSTEM chapter).

FGD-02F also includes a reference oscillator (*Reference*), which provides a reference frequency for temperature compensation of the sensor. The value of *Reference* is not affected by radiation dose.

The typical operation procedure of **FGD-02F** can be summarize as follows:

- 1. Sensor is charged to a predefined target value. This value should be as close as possible to the value of Reference.
- 2. The chip measures *Sensor* and *Reference*, each during a specific time window (see MEASUREMENT WINDOW SETTING chapter).
- 3. Sensor and Reference are read. The Sensor value drop is proportional to the radiation dose.
- 4. The value of *Reference* can be used for compensating temperature effects (see DATA POST-PROCESSING chapter).
- 5. If the sensor goes below a predefined threshold value, it is recharged to the original value (see RECHARGING SYSTEM chapter).

OPERATION DESCRIPTION

The operation of **FGD-02F** consists on alternating consecutive measurements of the *Sensor* and the *Reference*. The duration of the measurement is called *Measurement Window*. During the *Measurement Window*, **FGD-02F** counts *Sensor* and *Reference* pulses alternatively. The duration of a *Measurement Window* is equal for the *Sensor* and the *Reference* and it is configurable.

Reading Sensor and Reference

The values of the *Sensor* and the *Reference* are available at internal registers **F1S(17:0)** (*Sensor*) and **F1R(17:0)** (*Reference*) as 18-bit registers (see Table 2 and Table 3).

F1S(17:0) and F1R(17:0) can be read via SPI interface. They are updated after each

corresponding *Measurement Window* has elapsed. Therefore, it is recommended to wait always at least two Measurement Windows (plus an additional safety time of 10% *Measurement Window*) between two consecutive read commands.

FGD-02F can generate an interrupt signal at **NIRQ** pin after both *Sensor* and *Reference Measurement Windows* are finished. This is achieved by setting **MNREV** bit high (see Table 20). **NIRQ** is an active low signal.

DNEWR and **DNEWS** bits indicate if a new value is available since last individual bit check (see Table 7 and Table 8). They are cleared automatically after read



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 11/22

MEASUREMENT WINDOW SETTING

The Measurement Window is the total time that FGD-02F keeps counting Sensor and Reference pulses. Short windows allow higher measuring rates, while long windows can be used for filtering the measured values.

The *Measurement Window* is governed by pin **CK** and bit **ENGATE** (see Table 10):

- If ENGATE = 0 the Measurement Window is determined by a specific amount of pulses at CK pin.
- If ENGATE = 1 the Measurement Window determined by the duration of an external pulse at CK pin.

Measurement Window as amount of CK pulses

With bits **WINDOW(1:0)** there are four possible **CK** amount of pulses to be selected (see Table 9). E.g., if **WINDOW(1:0)** = 11, the *Measurement Window* will be active during 4096 pulses at pin **CK**.

Knowing the frequency from the signal at **CK**, it is easy to calculate the *Sensor* frequency:

$$Sensor\ Frecuency = \frac{F1S(17:0)}{Window\ Pulses\ amount} \times f(CK) \quad [Hz]$$

Similarly, the *Reference* frequency can be calculated.

Measurement Window gating at CK

In this configuration, the Measurement *Window* is active as long as **CK** pin is set high. Knowing the duration of **CK** pulse, *Sensor* frequency can be calculated:

Sensor Frecuency=
$$\frac{FIS(17:0)}{CK \text{ pulse duration}}$$
 [Hz]

Similarly, the *Reference* frequency can be calculated.

Figure 2 shows a timing diagram example of using *Measuring Window* gating at CK. CK must remain low a minimum time, tckl_{min} (*El. Char. No. 400*), between a *Sensor* and a *Reference* measurement. **DNEWS** (see Table 8) is set when *Sensor* measurement is finished, and **DNEWR** is set after a *Reference* measurement (see Table 7). If **MNREV** (see Table 20) is high, interrupt bit pin **NIRQ** will go low, signalling when the data should be read by serial communication. **DNEWS** and **DNEWR** are cleared automatically after read.

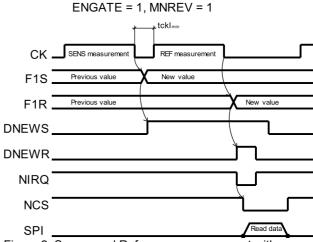


Figure 2: Sensor and Reference measurement with window gating

Count Overflow

F1S(17:0) and **F1R(17:0)** are 18-bit registers. If the selected *Measurement Window* is too long or the *Sensor* recharging value is too high, the registers might overflow. This event is flag through bits **F1SOVF** and **F1ROVF** (See Table 5 and Table 6).



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 12/22

SENSITIVITY CONFIGURATION

The *Sensor* offers two different sensitivity configurations, selected by bits **SENS(2:0**) (see Table 19):

- If SENS(2:0) = 100, Low Sensitivity configuration (LS) is selected. The recommended Sensor linear range goes from 110 kHz to 130 kHz. The sensitivity of the Sensor is lower, while the TID for triggering a recharge is higher.
- If SENS(2:0) = 001, High Sensitivity configuration (HS) is selected. The recommended Sensor linear range goes from 30 kHz to 50 kHz. The sensitivity of the Sensor is higher, while the TID for triggering a recharge is lower.

	kHz / Gy	Gy / Cycle
High Sensitivity	30	0.67
Low Sensitivity	5	4

Table 1: Sensitivity configuration

RECHARGING SYSTEM

The *Sensor* should be kept working within its linear range of 20 kHz. This range depends on the sensitivity configuration selected:

- High Sensitivity (HS): Typically from 30 kHz to 50 kHz
- Low Sensitivity (LS): Typically from 110 kHz to 130 kHz.

The *Reference* should be configured to the maximum value of the linear range. This is achieved using **REF(2:0)** bits (See Table 4).

If the *Sensor* value is discharged below the linear range, a recharging system allows recharging it back to the original value. This value should be as close as possible to the *Reference* value. The recharging system requires an external supply voltage between 15 V and 20 V at pin **VB** (*El. Char. No. 001*).

Two registers are available to define the upper and lower limits of the linear working range:

- TARGET(7:0) defines the maximum value (See Table 13). The recommended value is:
 - ∘ 50 kHz in HS
 - o 130 kHz in LS

- THRESHOLD(7:0) defines the minimum value (see Table 14). The recommended values is:
 - o 30 kHz in HS
 - o 110 kHz in LS

TARGET(7:0) and **THRESHOLD(7:0)** are 8-bit registers. To evaluate if the *Sensor* is out of the linear range defined by **TARGET(7:0)** and **THRESHOLD(7:0)**, they must be compared with the 8 MSBs of **F1S(17:0)**.

Configuring Target and Threshold

TARGET(7:0) and THRESHOLD(7:0) registers can be directly configured by asigning the following values:

- TARGET(7:0) = F1R(17:10) with maximum linear Sensor frequency (50 kHz or 130 kHz typically).
- THRESHOLD(7:0) = F1S(17:10) with minimum linear Sensor frequency (30 kHz or 110 kHz typically).



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 13/22

Alternatively, the following expressions can be used for configuring both registers, depending on the *Measurement Window* selected.

For Measurement Window as amount of CK pulses:

$$TARGET(7:0) = \frac{f(Sensor)}{f(CK)} \times \frac{Window\ Pulses\ amount}{1023}$$

where *f*(*Sensor*) is the maximum desired *Sensor* frequency (50 kHz or 130 kHz). The same expression can be used for **THRESHOLD**(**7:0**).

For Measurement Window as gating at CK:

$$TARGET(7:0) = f(Sensor)(Hz) \times \frac{CK \ pulse \ duration}{1023}$$

where *f*(*Sensor*) is the minimum desired *Sensor* frequency (30 kHz or 110 kHz). The same expression can be used for **THRESHOLD**(**7:0**).

There are two main recharging modes, depending on bits **CHMODE(1:0)** (see Table 11):

- Manual Recharge
- Automatic Recharge

Independent of the recharging mode, to enable recharges the global bit **ECH** must be set high (see Table 12).

Automatic Recharge

If **CHMODE(1:0)** = 01 the recharging system works automatically.

- When F1S(17:10) goes below THRESHOLD(7:0) a recharge start.
- When **F1S(17:10)** goes above **TARGET(7:0)** the recharge is stopped.

It is possible to track each recharge by setting **MNREV** bit low (see Table 20). **NIRQ** will be pulled low each time the *Sensor* is being recharged.

Bits RCHCNT(3:0) count the number of recharges carried out (see Table 18). They allow working with long periods between each data read. A read must be carried out before RCHCNT(3:0) reaches maximum value and it must be cleared manually by performing a write operation on address 0x01.

Bit **RCHEV** is a safety flag bit that indicates if a recharge is process (see Table 15).

Bit RCHRQ goes high if F1S(17:10) goes below THRESHOLD(7:0) (see Table 16). It can be used in combination with ECH bit to control when a recharge should be triggered in automatic mode.

Manual Recharge

In *Manual Recharge* the user controls the start and stop of the *Sensor* recharge. It can be controlled either by external pin or by internal bit:

- If **CHMODE(1:0)** = 10 the system recharges the *Sensor* while **FCH** bit is left high (see Table 17).
- If **CHMODE(1:0)** = 11 the system recharges the *Sensor* while **ENWR** pin is set high.

In *Manual Recharge* it is not necessary to use **TARGET(7:0)** and **THRESHOLD(7:0)**. To detect if the *Sensor* is within the linear range, **F1S(17:0)** must be polled. To relate the **F1S(17:0)** to the Sensor frequency the following expressions can be used, depending on the *Measurement Window* selected.

Measurement Window as amount of CK cycles:

$$FIS(17:0) = \frac{f(Sensor)}{f(CK)} \times Window Pulses amount$$

For Measurement Window as gating at CK:

$$FIS(17:0) = f(Sensor)(Hz) \times CK$$
 pulse duration

In manual recharge, the recharge counter is increased every time the user triggers a new recharge.



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 14/22

POWER SAVING MODE AND PASSIVE DETECTION

FGD-02F features two different modes for low and even zero power consumption: Power saving and Passive Detection Mode.

Both in Passive Detection and Power Saving Mode, the core of the sensor is still sensing and recording the received radiation dose.

For data reading, FGD-02F must be fully poweredone (Normal Operation). Once data reading operation is finished, it can be switched back to Passive Detection or Power Saving Mode-Mode.

Power Saving Mode

Bits **POWR(2:0)** allow reducing the power consumption by switching off some internal

functions, entering in a semi-iddle state (see Table 22):

- If **POWR(2:0)** = 000, power consumption is reduced.
- If POWR(2:0) = 111, power consumption is normal (Normal Operation)

Passive Detection Mode

FGDOS[®] can measure radiation dose with no supply voltage, acting as a passive radiation detector. Consumption of **FGD-02F** can be reduced to zero by switching off **VCC** and **VCCD** power supplies.

For data reading, **FGD-02F** must be powered-on. Once read, it can be switched back to passive detection mode.

INTERNAL TEMPERATURE MONITOR

FGD-02F includes an 8-bit temperature monitor with a range going from -40 °C to 125 °C and a resolution of 1 °C/LSB. The internal temperature can be obtained by reading **TEMP(7:0)** register, which is a read-only register (see Table 23).

Absolute read values may differ from one chip to another. An individual initial calibration of the temperature monitor is recommended.

The temperature monitor can be used to compensate temperature effects on the *Sensor*. The microcontroller can use a look-up table combined with the temperature value measured through **TEMP(7:0)** register.

SERIAL ID NUMBER

FGD-02F provides a 3 bytes-long unique individual serial number that can be read at address 0x10 to 0x12.

SERIAL PHERIPHERAL INTERFACE (SPI)

SPI slave interface

The SPI slave interface uses pins NCS, SCLK, MISO and MOSI. Pin NCS is the chip select pin and must be set lo by the SPI master in order to start communication. Pins MISO and MOSI are the data

communication lines and pin **SCLK** is the clock line generated by the SPI master (E.g. microcontroller).

The SPI protocol frames are shown in Figure 3. A communication frame consists of one address byte and at least one data byte. Bits 7:6 of the address



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 15/22

byte is the opcode used for selecting a read operation (set to "10") or a write (set to "01") operation. The remaining 6 bits are used for register addressing.

It is possible to transmit several bytes consecutively, if the **NCS** signal is not reset and **SCLK** keeps clocking. The address is internally incremented after each transmitted byte. Once the address reaches the last register (0x14h), it is reset back to 0x00.

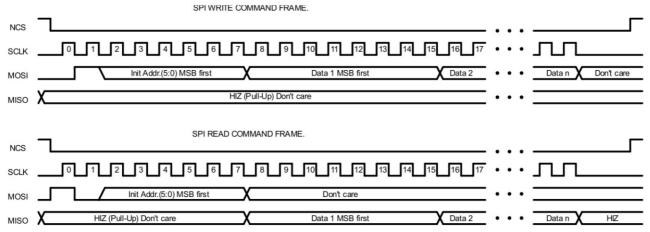


Figure 3: SPI Read and Write commands



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 16/22

DATA POST-PROCESSING

Background

The accuracy of **FGDOS**® is improved if the effects of operating temperature are compensated. This can be achieved by post-processing the sensor data through an external microcontroller or FPGA.

A single measurement of **FGDOS**® consists of a reference frequency (*FR*) and a sensor frequency (*FS*) pair. Both sensor and reference dependencies should be compensated to improve **FGDOS**® accuracy.

Typical values for FS and FR temperature dependence (see *El. Char. No. 209 and 302*).

Compensating for the FS and FR temperature dependence

FGDOS® has to be characterized in temperature after the sensor has been charged for the first time. This temperature characterization must be carried out under no radiation.

The relation of FS and FR is very linear with temperature variation. Assuming this linearity, the equation of a line can be extracted by measuring two pairs of FS and FR at different temperatures, T_{RT} and T_1 :

$$FS_{RAD0}(T_{RT}), FR_{RAD0}(T_{RT})$$

$$FS_{RAD0}(T_1), FR_{RAD0}(T_1)$$

where $FS_{\text{RADO}}(T_{\text{RT}})$ is the sensor frequency with no radiation at room temperature. This line is shown in Figure 4. The resulting equation is:

$$FS_{RAD0} = m \cdot FR_{RAD0} + a \tag{1}$$

Once the FS_{RAD0} is obtained, radiation can be applied to $FGDOS^{\otimes}$. When radiation RAD_1 is applied to the sensor, the line relating FS and FR is modified, but it can be assumed that the slope remains constant. Figure 5 shows the effect of applying radiation. When a pair of FR and FS is

measured under radiation RAD_1 and a random temperature T_3 , the following pair is obtained:

$$FS_{RAD1}(T_3), FR_{RAD1}(T_3)$$

With $FR_{RAD1}(T_3)$ and formula (1), $FS_{RAD0}(T_3)$ can be obtained. From Figure 4 it can be seen that:

$$FS_{RAD0}(T_3)$$
- $FS_{RAD0}(T_{RT})$ = $FR_{RAD1}(T_3)$ - $FS_{RAD1}(T_{RT})$

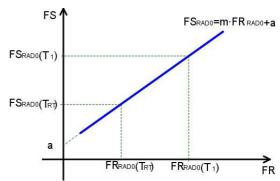


Figure 4: Variation of FS and FR with temperature

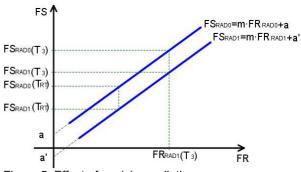


Figure 5: Effect of applying radiation

The radiation increase with respect to $\mathsf{FS}_\mathsf{RAD0}(\mathsf{T}_\mathsf{RT})$ is therefore:

$$FS_{RAD1}(T_{RT}) = FR_{RAD1}(T_3) - FS_{RAD0}(T_3)$$

This value is temperature compensated and is given in frequency. Applying the Frequency Sensitivity factor (*El. Char. No. 202*), the radiation value in Gy is obtained.



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 17/22

This compensation technique assumes linear behavior of the relation of FR and FS with temperature, as well as constant slope of this relation when radiation is applied.

This is a possible approach, however it is possible to improve even more the accuracy if these

assumptions are not considered and instead a look-up table is used for temperature compensation.



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 18/22

REGISTER DESCRIPTION

Sensor and Reference

F1S	Addr. 0x06/07/08	Bits 17:0	R 0x0000
0x00000	Minimum sens	sor counter val	ue
0x3FFFF	Maximum sen	sor counter va	lue

Table 2: Sensor Counter

F1R	Addr. 0x03/04/05	Bits 17:0	R 0x0000
0x00000h	Minimum refe	rence counter	value
0x3FFFFh	Maximum refe	rence counter	value

Table 3: Reference Counter

REF	Addr. 0x0B	Bits 6:4	R/W 000
000	Reference se	t to minimum	
111	Reference se	t to maximum	

Table 4: Reference frequency configuration

F1ROVF	Addr. 0x05 Bit 2	R 0
0	No Reference overflow	
1	Reference overflow	

Table 5: Reference counter overflow

F1SOVF Addr. 0x08 Bit 2 R 0 0 No Sensor overflow 1 Sensor overflow

Table 6: Sensor counter overflow

DNEWR	Addr. 0x05 Bit 3	R 0
0	No new reference data is ready	
1	New reference data is ready	

Table 7: Reference data new value

DNEWS	Addr. 0x08 Bit 3	R 0
0	No new sensor data is ready	
1	New sensor data is ready	

Table 8: Sensor data new value

Measurement Window bits

WINDOW	Addr. 0x0B Bit 3:2	R/W 00
00	32,768 CK pulses per window	
01	16,384 CK pulses per window	
10	8,192 CK pulses per window	
11	4,096 CK pulses per window	

Table 9: Window length selection

ENGATE	Addr. 0x0E Bit 0	R/W 0
0	Meas. Window by counts at CK	
1	Meas. Window by gating at CK	

Table 10: Enable window gating

Recharging System bits

CHMODE	Addr. 0x0D	Bits 1:0	R/V	V 00
00	Recharging of	disabled		
01	Automatic re	harging m	node	
10	Manual recha	arging cor	ntrolled by F	СН
11	Manual red ENWR pin	charging	controlled	by

Table 11: Charge Mode Selection

1	ECH	Addr. 0x0D	Bit 6	R/W 0
	0	Recharge not	allowed	
	1	Recharge allo	wed	

Table 12: Enable Recharging

TARGET	Addr. 0x09	Bits 7:0	R/W 0x00
0x00	Minimum Fred	quency	
0xFF	Maximum Fre	quency	

Table 13: Upper level target frequency

THRES	Addr. 0x0A	Bits 7:0	R/W 0x00
0x00	Minimum Fre	quency	
0xFF	Maximum Fre	equency	

Table 14: Lower level threshold frequency

RCHEV	Addr. 0x01 Bit 7	R 0
0	Recharge not in progress	
1	Recharge in progress	

Table 15: Recharge event flag



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 19/22

RCHRQ	Addr. 0x01 Bit 6	R 0
0	Recharge is not requested	
1	Recharge is requested	

Table 16: Recharge request

FCH	Addr. 0x0C Bit 7	R/W 0
	FG charging stopped	
1	FG charging started	

Table 17: Force charge in Manual Recharging mode

RCHCNT(3:0)	Addr. 0x01	Bit 3:0	R/W 0000
0000	No recharges	since last	clear
1111	At least 15 rech. since last clear		

Table 18: Recharge counter

Sensitivity Configuration

SENS(2:0)	Addr. 0x0C Bit 2:0	R/W 000
XXX	Reserved	
001	High Sensitivity Sele	cted
XXX	Reserved	
100	Low Sensitivity Select	cted
XXX	Reserved	

Table 19: Sensitivity Configuration

Interrupt request

MNREV	Addr. 0x0E	Bit 6	R/W 0
0	NIRQ signals	recharge in proce	ess
1	NIRQ signals	measurement rea	adv

Table 20: Measurement / Recharge indicator

NIRQOC	Addr. 0x0E Bit 1	R/W 0
0	NIRQ interruption push-pull	
1	NIRQ interruption open collector	r

Table 21: Interrupt output

Power Saving and Temperature Monitor

POWR(2:0)	Addr. 0x0C Bits 5:3 R/W 111		
XXX	Reserved		
111	Normal Operation		
XXX	Reserved		
000	Power saving Mode		
XXXX	Reserved		

Table 22: Power saving

TEMP	Addr. 0x00	Bits 7:0	R 0x00
0x00	Minimum tem	perature value	
0xFF	Maximum ten	nperature value	

Table 23: Temperature monitor

Serial Number and Chip Version

SN	Addr. 0x10/11/12	Bits 23:0	R 0x000000
0x000000 0xFFFFF	Minimum se Maximum se		

Table 24: Serial Number

CHIPID	Addr. 0x13	Bits 7:0	R 0x01
0x01	FGD-02F ve	rsion 1	

Table 25: Chip Identification Number



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 20/22

REGISTER MAP

OVERVIEV	٧							
Addr	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x00 R				TEM	P(7:0)			
0x01 R	RECHEV	RECHRQ	0	0		RCHC	NT(3:0)	
0x02				Not imp	lemented			
0x03 R				F1R	2(7:0)			
0x04 R				F1R	(15:8)			
0x05 R	0	1	0	1	DNEWR	F1ROVF	F1R(17:16)
0x06 R				F18	6(7:0)			
0x07 R				F1S	(15:8)			
0x08 R	1	0	1	0	DNEWS	F1SOVF	F1S(17:16)
0x09	TARGET(7:0)							
0x0A				THRESH	IOLD(7:0)			
0x0B	0*		REF(2:0)		WINDO	OW(1:0)	0*	0*
0x0C	FCH	1*		POWR(2:0))		SENS(2:0)	
0x0D	0*	ECH	0*	0*	0*	0*	СНМО	DE(1:0)
0x0E	0*	MNREV	1*	1*	0*	0*	NIRQOC	ENGATE
0x0F				Not imp	lemented			
0x10	SN(7:0)							
0x11	SN(15:8)							
0x12	SN(23:15)							
0x13 R		CHIPID(7:0)						
0x14	0	0	0	0		Rese	erved	

R: Read-only register

(*): Reserved. Must be set to specified value



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 21/22

DESIGN REVIEW: Notes On Chip Functions

FGD-02F Z			
No.	Function, Parameter/Code	Description and Application Notes	
1	CHIPID	For FGD-02F chip releases see Table 25	

DATASHEET REVISION HISTORY

Rel.	Rel. Date	Chapter	Modification	Page
A0.5		Sensitivity	LS and HS frequency ranges are typical recommended values	
A0.6		All	Minor errors on tables	
A0.7		Design Review	Added info regarding Chip release version	21/21
A0.8		Ordering Information	EVAL USB RTC	21/21
A0.9		Disclaimer		21/22
A0.10		Quick Set-up and Operation example	Addr. $0x0C = b01111001 = 0x79$	8/22
A0.11	07/11/19	ELECTRICAL CHAR.	004; I(VCC) and 006; I(VCCD)	5/22
		Datasheet	Changed from "Confidential" to "Preliminary"	
		Datasheet	Changed contact info	
		OREDERING INFO.	Removed FGD-02F_Z TC	22/22

SOLDERING CONSIDERATIONS

The temperature applied to FGD-02F throughout the soldering process can lead to charges recombination on FGDOS® sensor. Consequently,

it is recommended to trigger a new charging process after soldering.



Floating Gate Dosimeter (FGDOS®)



Target Specification. Preliminary

Rev. A0.11 Page 22/22

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ORDERING INFORMATION

Product	Description
FGD-02F_Z	FGD-02F_Z sensor, non-characterized, QFN32
FGD-02F_Z RTC	FGD-02F_Z sensor, radiation and temperature characterized, QFN32
FGD-02F EVAL RTC	Evaluation board with FGD-02F_Z sensor, radiation and temperature characterized

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